



Accurate Waveform Modeling using SVD with Applications to Timing Analysis

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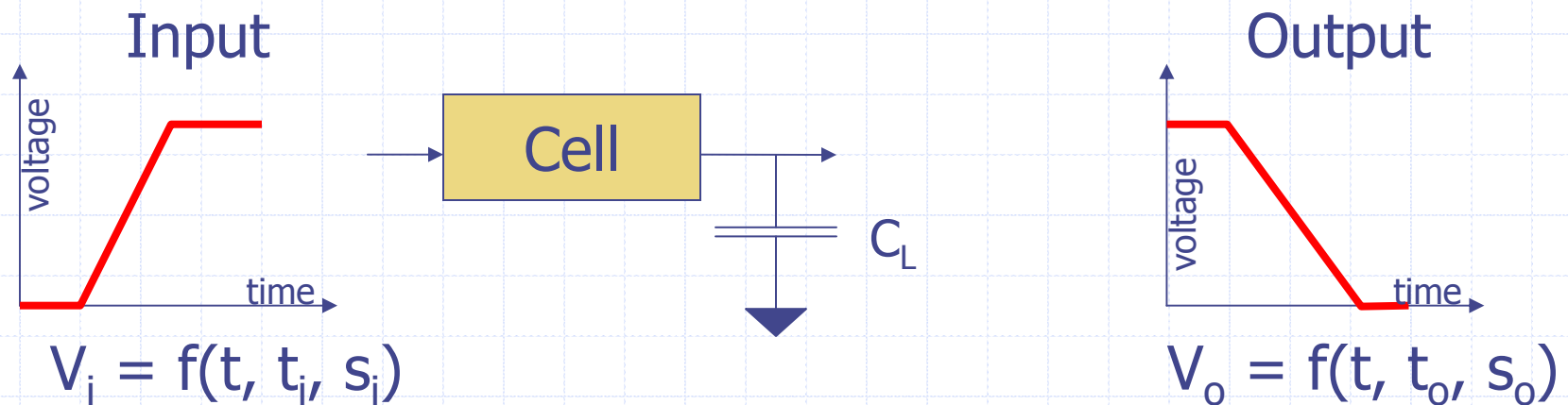
Motivation

- ◆ On-chip waveforms do not look like “ramps”
 - Devices are operating in more complex regimes and do not at all look like current sources
 - Loads, both interconnect as well as gate inputs, are resistive and non-linear

- ◆ We persist in trying to fit an outdated waveform model onto far more complicated behaviors
 - Applications like Statistical Static Timing Analysis (SSTA) require accurate modeling
 - Model inaccuracy must be \ll expected variability to reliably estimate performance variability

Ramp Based Timing Model

- ◆ Expressed in terms of a Ramp approximation of input and output waveforms.
 - Arrival time, Slope and Load Capacitance



Timing Model: $[t_o, s_o] = F([t_i, s_i], C_L)$

Source of Error in Timing Models

1. Inability of the waveform function (ramp) to fit the real waveform.
2. Estimation of a complex load by a single capacitance.
3. Lack of complete modeling support (coupling noise, multiple input switching etc...).

- ◆ We are focusing in this paper on the first two sources of error:
 - Part 1: π -model for interconnect
 - Part 2: Accurate waveform modeling

Part 1: Better Load Modeling

◆ Benefits:

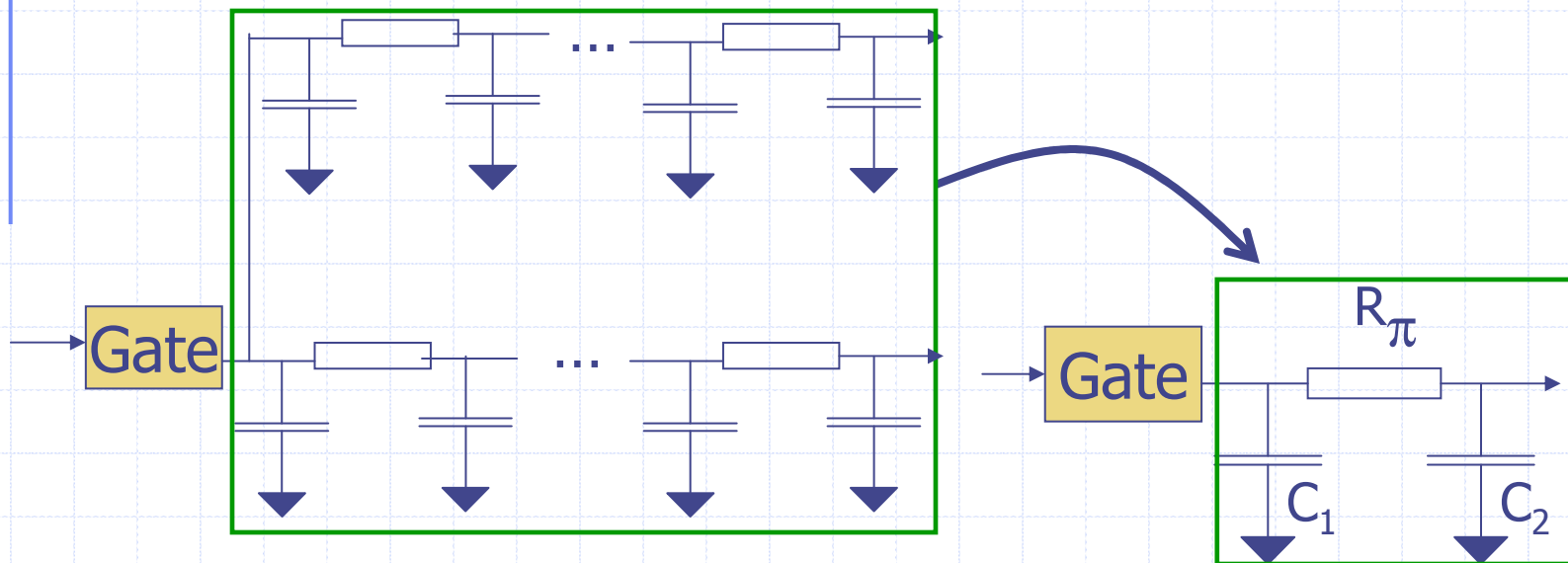
- A π model of the load is clearly a better representation than a single capacitor.
- We did not do a complete study to exactly quantify the improvement achieved.

◆ Costs:

- Modeling a gate's behavior as a function of a π load means we have 2 more variables to vary.
- When using traditional (e.g. full factorial) experiment designs to create the timing models, adding 2 variables can be quite costly.

Existing Work on π -Models

- ◆ O'Brien and Savarino developed an algorithm for reducing an RC tree to a driving-point π -model.



Solving the Dimensionality Problem

- ◆ A naïve implementation of a gate model builder may use a full factorial design, resulting in an exponential number of simulation vs. modeling variables.
- ◆ We use Latin-Hypercube Sampling, a well established statistical sampling technique instead.
 - Number of simulation \sim linear in number of modeling variables.

Part 2: Accurate Waveform Modeling

History:

◆ Heuristic models

- Equivalent waveform model [Hashimoto, ICCAD '03]
- Weibull distribution [Amin, ICCAD '03]

◆ Change of basis models

- Model current not voltage, CSM [Amin, DAC '06]

◆ Data based models

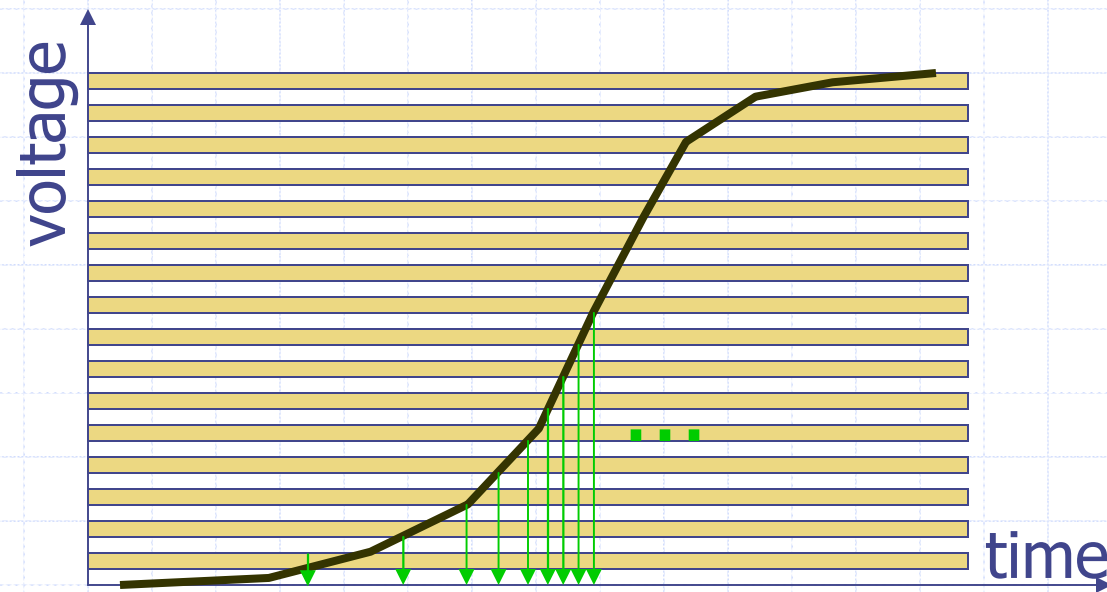
- Basis decomposition [Jain, ICCAD '05]
- PCA based approach [Nassif, TAU '04]

Our approach: extend the PCA approach

- Use more appropriate SVD instead of PCA
- Generate waveform model based on complete library
- Demonstrate application to interconnect as well

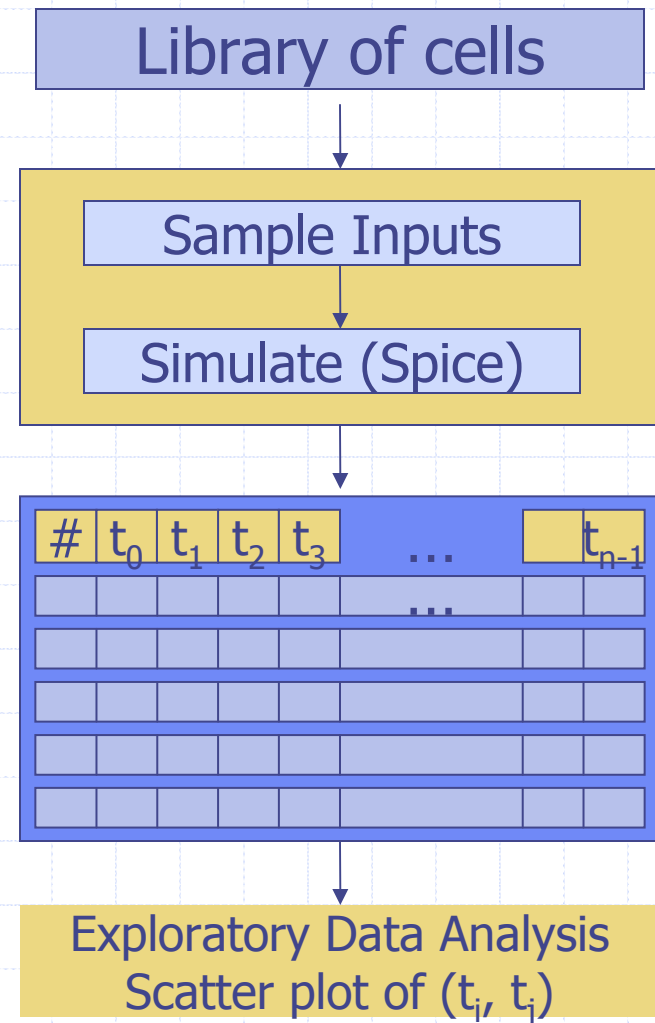
Data based model

- ◆ Divide $[0 \dots V_{dd}]$ into n intervals.
- ◆ Measure t_0 through t_{n-1} for waveforms of interest.
 - $t_i =$ time at which waveform crosses $V_{dd} \times (i/(n-1))$

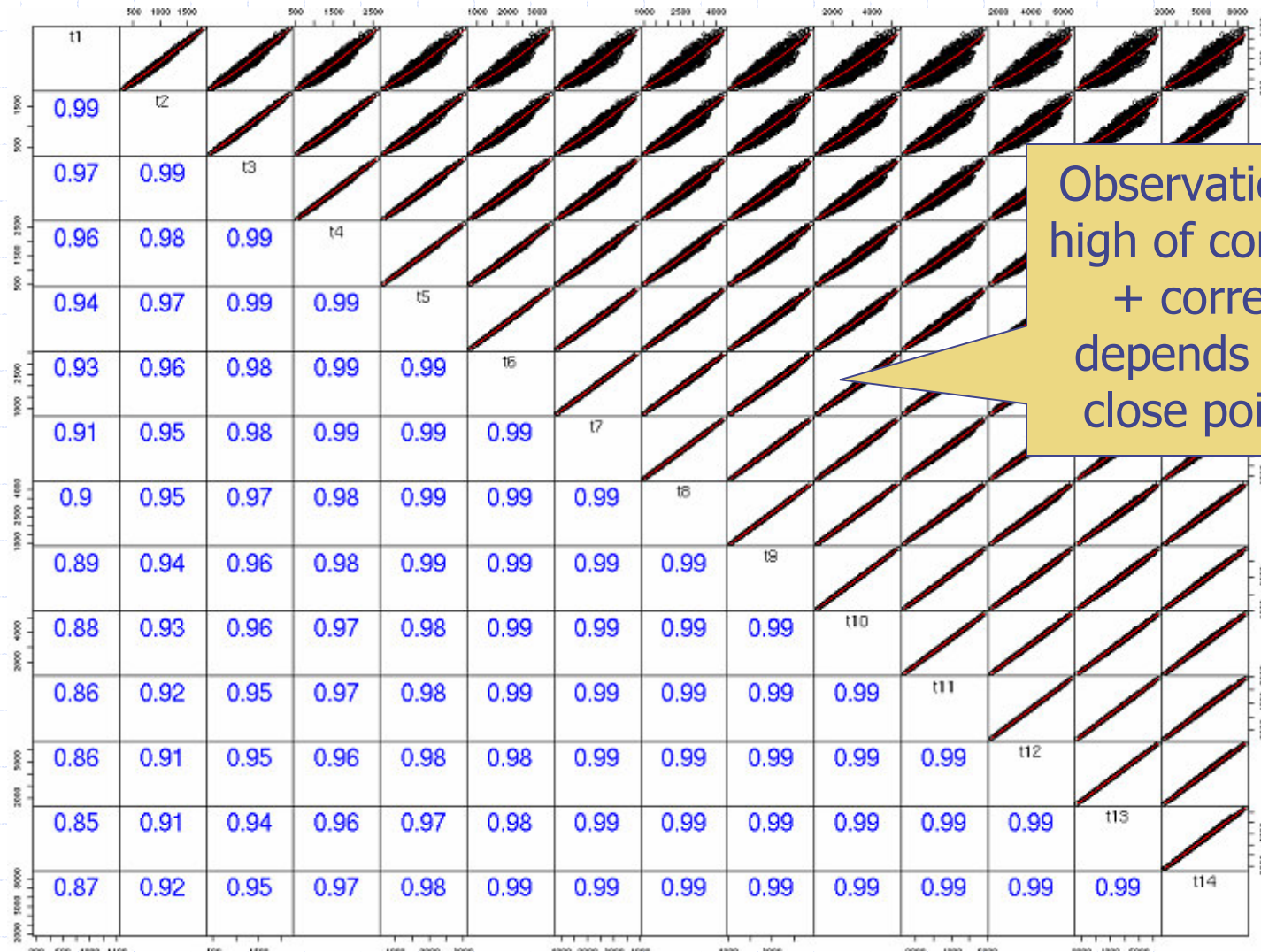


Intuition for our work

- ◆ The time/voltage (t_i, v_i) pairs that define a waveform are not independent of each other
- ◆ To verify this we analyzed waveforms obtained from various cells in the library under varying input and interconnect load conditions
 - We expect the crossing times (t_i) of these waveforms to be inter-related.



Scatter plot of (t_i, t_j) pairs



Observation: Very high of correlation, + correlation depends on how close points are

Comments on Crossing Time Stats

- ◆ The crossing times (t_i) are obviously not independent of each other
 - Strong correlation across all the crossing times
 - Then t_i can be expressed as a function of a smaller number of independent variables
- ◆ How to find a smaller subset of independent variables?
 - Previous work used PCA, which works best when the distribution of the t_i is Gaussian (not the case in general)
- ◆ We use an alternative dimension reduction technique, Singular value decomposition (SVD)
 - Designed for the more general case where the t_i are simply linearly related.

Singular Value Decomposition (SVD)

- ◆ SVD of a matrix \mathbf{T} is given by

$$\mathbf{T} = \mathbf{U} \mathbf{\Sigma} \mathbf{V}'$$

- ◆ \mathbf{U} – *orthonormal* basis for columns in \mathbf{T}

- ◆ \mathbf{V} – *orthonormal* basis for the rows in \mathbf{T}

- Thus the *basis for waveforms*
- Note that the basis are obtained from the *data*
- Thus, data speaks for itself (no assumptions needed)

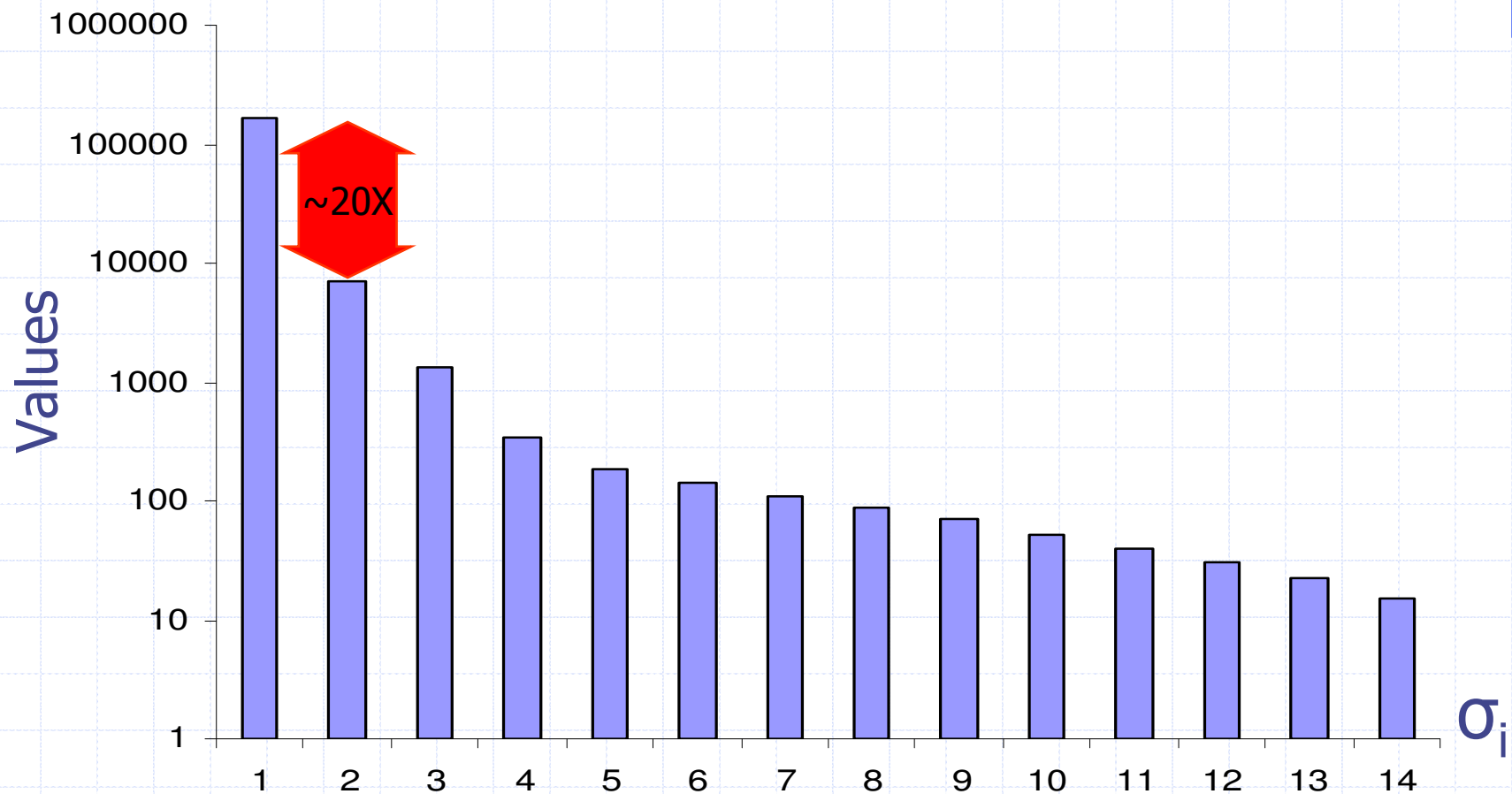
- ◆ $\mathbf{\Sigma}$ – diagonal matrix contains singular values

- Singular values are ordered in a *non-decreasing* order
- Singular values σ_i “weighs” the basis columns $\mathbf{V}_{\cdot i}$
- First few basis are sufficient to capture the data (i.e. the waveform) accurately

#	t_0	t_1	t_2	t_3	...	t_{n-1}
					...	

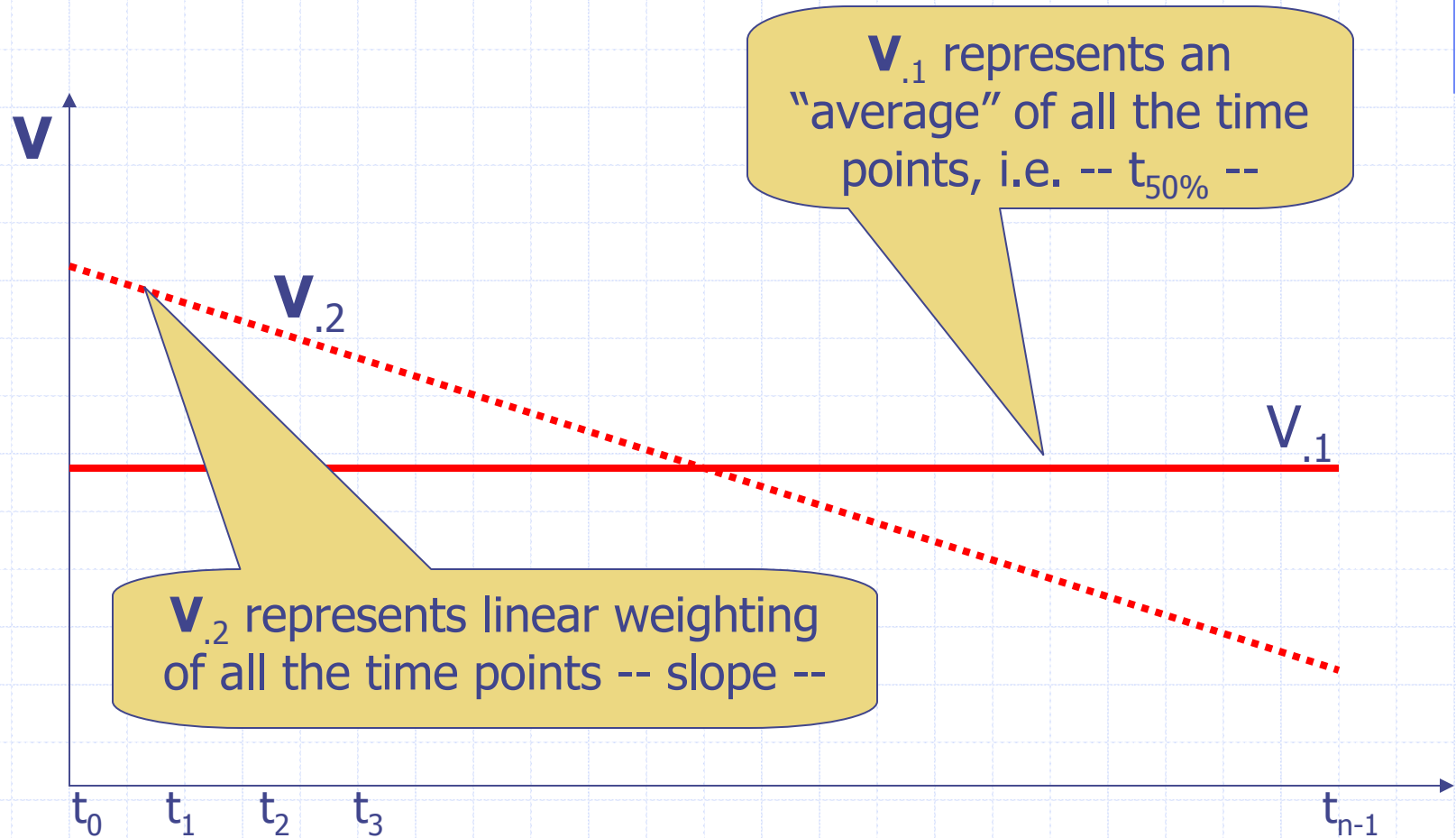
Singular values of a Matrix

- ◆ A plot of singular values of \mathbf{T}
 - Only the first few dominate



Interpreting the basis vectors (\mathbf{V})

- ◆ Each of the columns of \mathbf{V} represents a weighted sum of the times t_i



Example

- ◆ If the n time points of a *waveform* are represented as pairs (voltage _{j} , t_j)

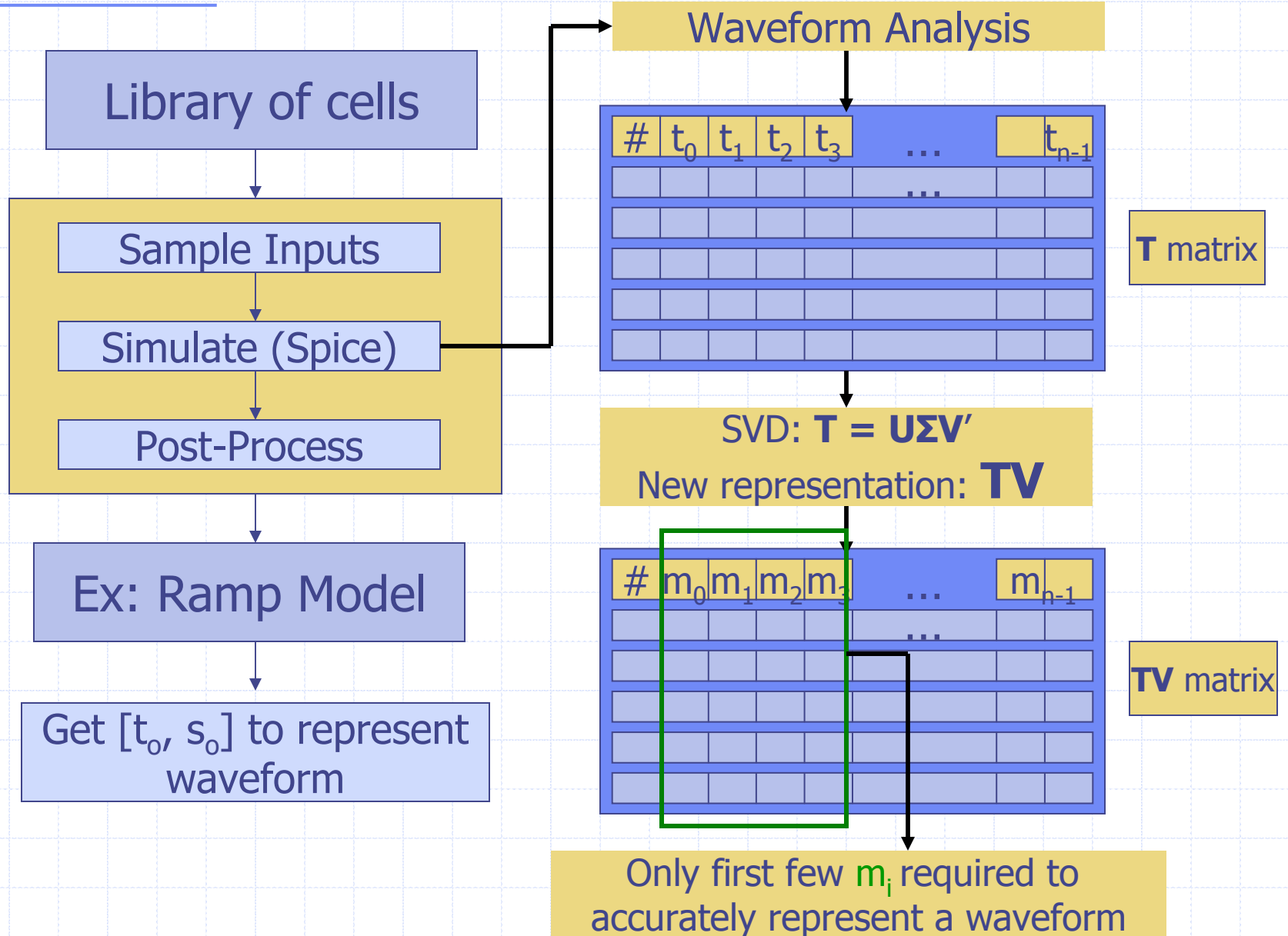
Voltage	0	1/13	...	12/13	1
t [ps]	599	632	...	788	802

- ◆ Consider $\mathbf{V}_{.2}$, which we interpreted as \sim slope in the previous slide

$\mathbf{V}_{.2}$	$V_{1,2}$	$V_{2,2}$...	$V_{13,2}$	$V_{14,2}$
value	-0.52	-0.41	...	0.31	0.39

- ◆ The value in the new basis is given by dot product $\langle \mathbf{t}, \mathbf{V}_{.2} \rangle = -202.31$, which approximates the slope.
 - We call this dot product a *moment* (m_2)

Summary of SVD Analysis



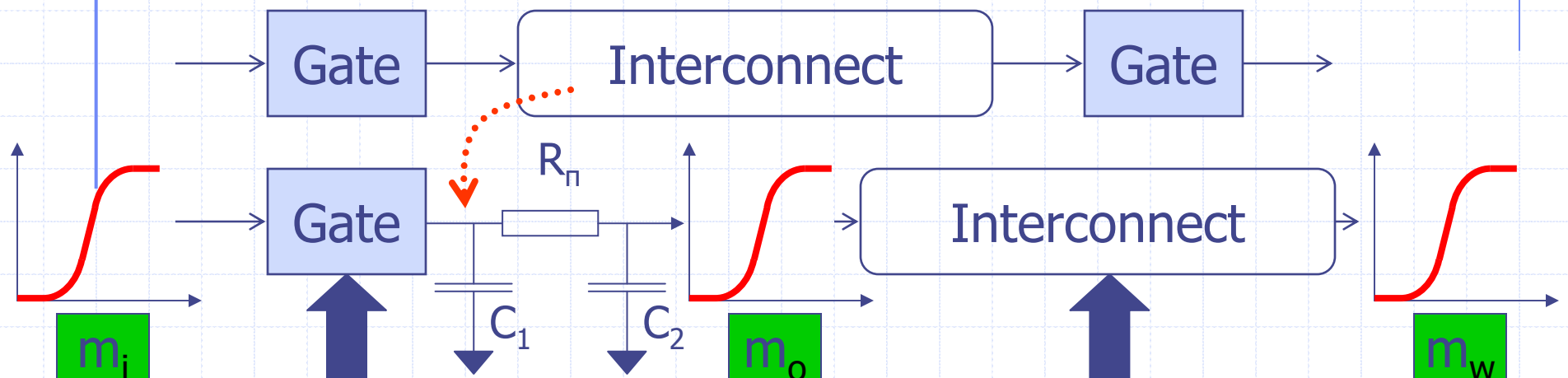
What Does This All Mean?

- ◆ It is possible to sample, simulate and analyze the set of waveforms that a *library of cells* would produce
 - From this we can determine precisely how many independent variables are required in order to represent waveforms with a specified accuracy.
 - When we analyze the entire library we might need more independent variables than for a given cell.

- ◆ Once the independent variables are selected, we also get a transformation that allows us to go from the independent variables to the waveform
 - So the complete waveform can be readily re-generated from the values of those variables.

SVD Models in STA

- ◆ We need to propagate waveforms through gates and through wires.



Gate Timing Model:
 $[m_o] = F_G([m_i], C_L)$

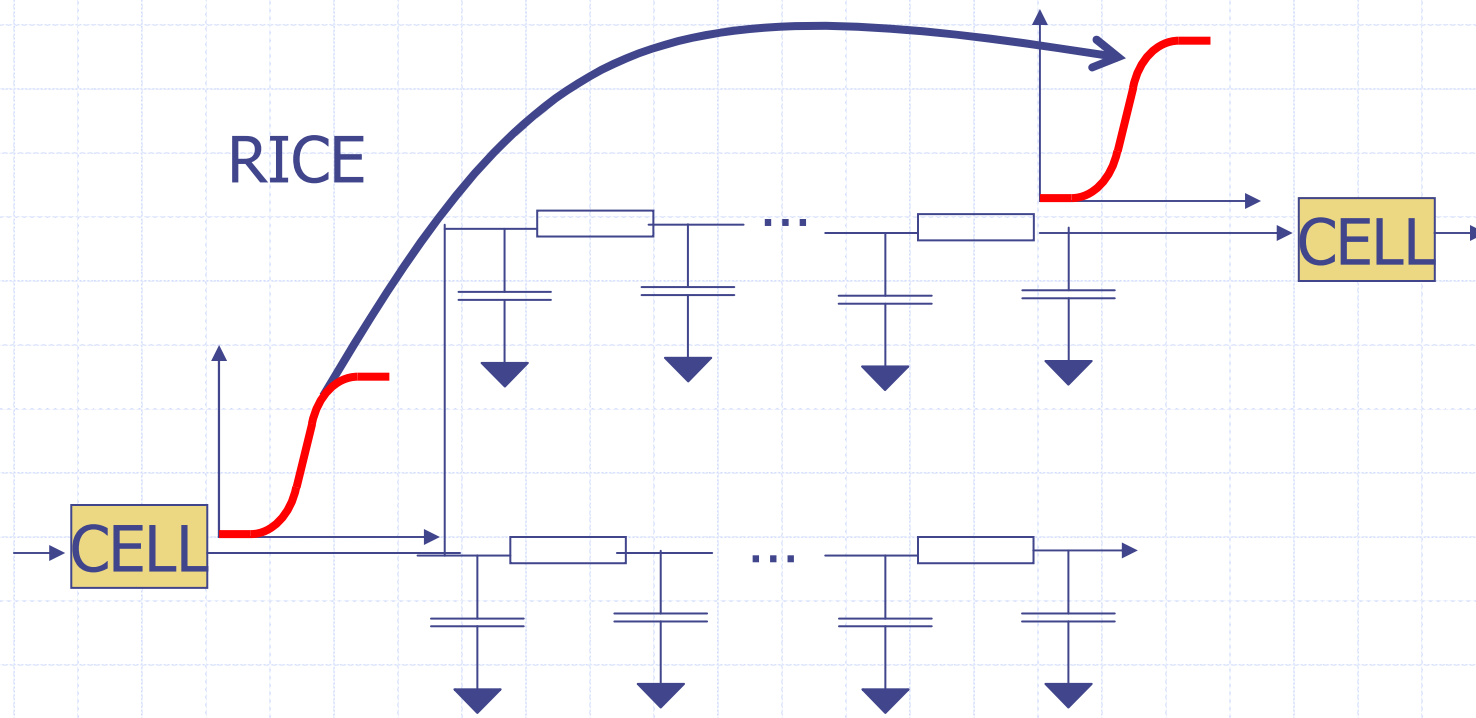
Implemented using
 enhanced timing models

Wire Timing Model:
 $[m_w] = F_W([m_o], C_L)$

We will explore this next

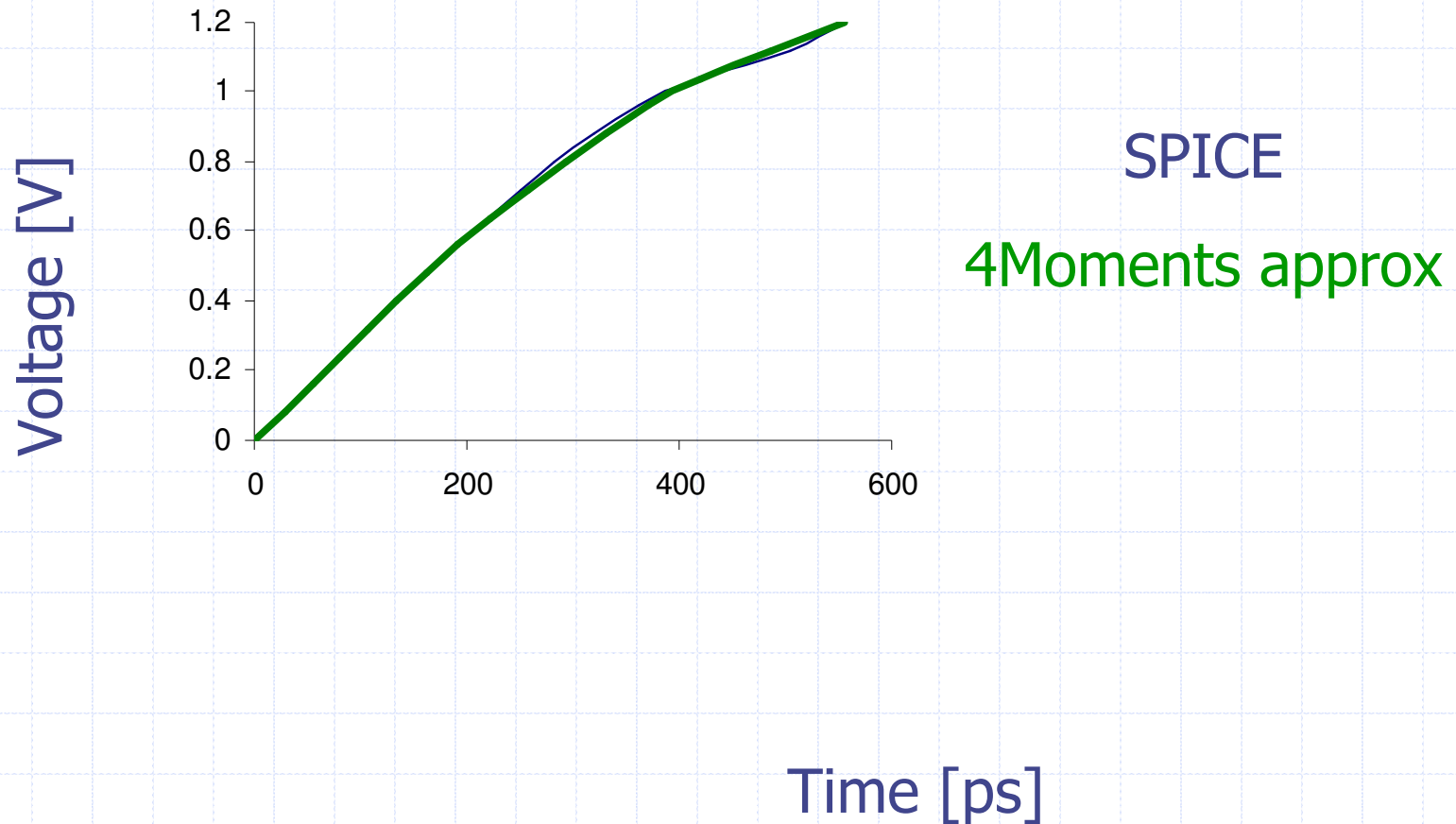
SVD + STA (Interconnects)

- ◆ We will still use RICE to propagate the waveform
- ◆ We will use the SVD transforms to convert back and forth between real waveform representation (voltage vs. time) and moments!



Waveform recovery from moments

- ◆ Waveform recovered at the far end of the interconnect of an inverter



Where Is This Model Needed

- ◆ Any time that an interface between the analog and digital world is required.
 - **Input/Output from wire loads.**
- ◆ Any time that knowledge of the waveform details is desired
 - SSTA, where model inaccuracy must be « expected variability to reliably estimate performance variability
 - Another Example: Estimating I_{DD} in power grid simulation
 - ◆ A ramp waveform estimate not useful for since it makes the current look like a step

Features of our model

- ◆ Precisely quantify the error we commit in modeling waveforms
- ◆ A model which can gracefully expand to model additional effects
 - Resistive wires, process variations, ...
- ◆ A model which is a natural extension of the existing models
 - Allows us to use existing models where possible

Conclusions

- ◆ Advent of SSTA is causing a re-examination of how cell delay models are generated.
 - Additional dependencies are required.
 - More accuracy is needed.

- ◆ Empirical enhancements are costly in development time.

- ◆ A data-driven approach which re-uses existing data to drive improvement has the best chance of success.